

Design Of Computer Program For Assessment Of The Impact Of Satellite Antenna Pointing Error On The Effective Parabolic Antenna Gain And Aperture Diameter

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Abstract— The design of computer program for assessment of the impact of satellite antenna pointing error on the effective parabolic antenna gain and aperture diameter is presented. The program also implemented parametric analysis of the effect of signal frequency on the antenna gain, effective aperture diameter for different antenna pointing errors. The mathematical expressions for the computations were presented along with the design of the program algorithm. The program was implemented in Visual Basic for Application. Sample numerical examples were presented with antenna efficiency of 0.65, parabolic antenna diameter of 1.2 m, antenna pointing error range of 0 ° to 0.5° and frequency range of 100 MHz to 2100 MHz. The program results were presented in tables and graphs with trend line fitted on the graphs. In all, the program results provides comprehensive analysis of the impact of the antenna pointing error and frequency on the key parameters of the parabolic antenna.

Keywords — Antenna Gain , Pointing Error , Aperture Diameter, Satellite Antenna, Parabolic Antenna

1. Introduction

Satellite communications operates mainly in the microwave frequency band which works best under line-of-sight conditions [1,2,3,4,5,6,7,8,9,10,11]. In this case, there should be clear line-of-sight between the transmitter antenna and the receiver antenna [12,13, 14,15, 16,17, 18, 19, 20]. When parabolic antenna is used, the parabolic antenna will focus and direct the transmitted signal along the line-of-sight to the receiving parabolic antenna [21,22,23,24,25,26,27,28]. Accurate alignment of the transmitting and receiving antennas is essential for maximum signal power reception at the receiver [29,30,31,32,33,34].

In some cases, there is antenna pointing error which cause some losses in the effective antenna gain

[35,36,37,38,39,40,41,42,43,44,45]. Hence, the actual antenna gain obtained when there is no pointing error is different from the antenna gain when there is pointing error. The difference in antenna gain also means that the effective parabolic antenna diameter is also affected. Hence, this paper is mean to present the development and application of a computer program for assessing the impact of pointing error on the effective antenna gain of parabolic antenna in the presence of pointing error, the effective antenna diameter with pointing error and other relevant parameters associated with antenna pointing error. The detailed algorithm for the program is presented along with numerical examples showing the applicability of the program.

2. Methodology

2.1 The mathematical expressions parametric analysis of impact of satellite antenna pointing error

When a parabolic antenna is designed for a signal with λ wavelength in meters, if the parabolic antenna has d_{mx} diameter in meters and efficiency of η , then, the antenna gain, $G_{mx(dB)}$ in dB is given by;

$$G_{mx(dB)} = 10 \text{Log} \left((\pi^2) \eta \left(\frac{d_{mx}}{\lambda} \right)^2 \right) \quad (1)$$

$$G_{mx(dB)} = 9.942997 + 10 \text{Log}(\eta) + 20 \text{Log} \left(\frac{d_{mx}}{\lambda} \right) \quad (2)$$

The θ_{3dB} antenna beam width is expressed as;

$$\theta_{3dB} = 70 \left(\frac{\lambda}{d_{mx}} \right) = 70 \left(\frac{3 \times 10^8}{(f) d_{mx}} \right) \quad (3)$$

Where the f is given in Hz. If there is antenna pointing error of θ , then , the pointing loss, L_θ of the antenna is expressed as;

$$L_\theta = 12 \left(\frac{\theta}{\theta_{3dB}} \right)^2 = 12 \left(\frac{\theta}{70} \left(\frac{\lambda}{d_{mx}} \right) \right)^2 \quad (4)$$

The effective antenna gain, $G_e(dB)$ is computed as;

$$G_e(dB) = G_{mx(dB)} - L_\theta \quad (5)$$

$$G_e(dB) = 10 \text{Log} \left((\pi^2) \eta \left(\frac{\lambda}{d_{mx}} \right)^2 \right) - 12 \left(\frac{\theta}{70} \left(\frac{\lambda}{d_{mx}} \right) \right)^2 \quad (6)$$

The diameter, D_e for the effective antenna gain, $G_{e(dB)}$ is computed as;

$$D_e = \lambda \sqrt{\frac{10^{\left(\frac{G_{e(dB)}}{10}\right)}}{(\pi^2)\eta}} \quad (7)$$

Change in antenna diameter, Δd is given as;

$$\Delta d = d_{mx} - d_e \quad (8)$$

Algorithm 1 The procedure for the Module Main ()

Algorithm 1 The procedure for the Module Main ()

Module Main ()

```

1:Input SelectParam /* Select the parametric to be used for the analysis ; Integer 1 to 5
    '1' for parametric analysis using ( $\lambda$ )
    '2' for parametric analysis using (f)
    '3' for parametric analysis using ( $\eta$ )
    '4' for parametric analysis using ( $d_{mx}$ )
    '5' for parametric analysis using ( $\theta$ )
*/
2:If SelectParam = 1 Then
3:  Module 1 ParAnal_using_ $\lambda$ 
4: Elseif SelectParam = 2 Then
5:  Module 2 ParAnal_using_f
6: Elseif SelectParam = 3 Then
7:  Module 3 ParAnal_using_ $\eta$ 
8: Elseif SelectParam = 4 Then
9:  Module 4 ParAnal_using_ $d_{mx}$ 
10: Else
11: Module 5 ParAnal_using_ $\theta$ 
12: EndIf
13: End Module main

```

Algorithm 2 The procedure for the Module 1 ParAnal_using_ λ ()

Algorithm 2 The procedure for the Module 1 ParAnal_using_ λ ()

Module 1 ParAnal_using_ λ ()

```

1: Input  $\lambda_L$  // lower value of wavelength
2: Input  $\lambda_H$  // higher value of wavelength
3: Input  $n_x$  // number of steps to cover the range  $\lambda_L$  to  $\lambda_H$ 
4: Declare the following arrays  $\lambda[n_x]$ ,  $G_{mx(dB)}[n_x]$ ,  $\theta_{3dB}[n_x]$ ,  $L_\theta[X]$ ,  $G_{e(dB)}[n_x]$ ,  $d_e[n_x]$ ,
     $\Delta d[n_x]$ 
5: Input  $\eta$ 
6: Input  $d_{mx}$ 
7: Input  $\theta$ 
8: Declare the following arrays  $\eta[n_x]$ ,  $d_{mx}[n_x]$ ,  $\theta[n_x]$ 
9: For  $X = 1$ ,  $X$  (  $n_x$ ,  $X = X + 1$ 
    10:  $\lambda[X] = \lambda \left( \frac{\lambda_H - \lambda_L}{n_x} \right)$ 
    11:  $\eta[X] = \eta$ 
    12:  $d_{mx}[X] = d_{mx}$ 

```

13: $\theta[X] = \theta$

14: Compute $G_{mx(dB)}[X] = 10 \text{Log} \left((\pi^2) \eta[X] \left(\frac{d_{mx}[X]}{\lambda} \right)^2 \right)$ using equation (1)

15: Compute $\theta_{3dB}[X] = 70 \left(\frac{\lambda[X]}{d_{mx}[X]} \right)$ using equation (3)

16: Compute $L_\theta[X] = 12 \left(\frac{\theta[X]}{70} \left(\frac{\lambda[X]}{d_{mx}[X]} \right) \right)^2$ using equation (4)

17: Compute $G_{e(dB)}[X] = G_{mx(dB)}[X] - L_\theta[X]$ using equation (5)

18: Compute $d_e[X] = \lambda[X] \sqrt{\frac{10 \left(\frac{G_{e(dB)}[X]}{10} \right)}{(\pi^2) \eta[X]}}$ using equation (7)

19: Compute $\Delta d[X] = d_{mx}[X] - d_e[X]$ using equation (8)

20: Output $\lambda[n_x]$, $\eta[X]$, $d_{mx}[X]$, $\theta[X]$, $G_{mx(dB)}[n_x]$, $\theta_{3dB}[n_x]$, $L_\theta[X]$, $G_{e(dB)}[n_x]$, $d_e[n_x]$, $\Delta d[n_x]$

21: Next For Loop

22: End For Loop

23: End Module 1 ParAnal_using_λ ()

Algorithm 3 The procedure for the Module 2 ParAnal using f()

Algorithm 3 The procedure for the Module 2 ParAnal_using_f()

Module 2 ParAnal_using_f()

1: Input f_L // lower value of frequency in Hz

2: Input f_H // higher value of frequency in Hz

3: Input n_x // number of steps to cover the range f_L to f_H

4: Declare the following arrays $\lambda[n_x]$, $G_{mx(dB)}[n_x]$, $\theta_{3dB}[n_x]$, $L_\theta[X]$, $G_{e(dB)}[n_x]$, $d_e[n_x]$, $\Delta d[n_x]$

5: Input η

6: Input d_{mx}

7: Input θ

8: Declare the following arrays $\eta[n_x]$, $d_{mx}[n_x]$, $\theta[n_x]$

9: For $X = 1$, $X < n_x$, $X = X + 1$

10: $f[X] = X \left(\frac{f_H - f_L}{n_x} \right)$

11: $\lambda[X] = \left(\frac{3 \times 10^8}{f[X]} \right)$

12: $\eta[X] = \eta$

13: $d_{mx}[X] = d_{mx}$

14: $\theta[X] = \theta$

15: Compute $G_{mx(dB)}[X] = 10 \text{Log} \left((\pi^2) \eta[X] \left(\frac{d_{mx}[X]}{\lambda} \right)^2 \right)$ using equation (1)

16: Compute $\theta_{3dB}[X] = 70 \left(\frac{\lambda[X]}{d_{mx}[X]} \right)$ using equation (3)

17: Compute $L_\theta[X] = 12 \left(\frac{\theta[X]}{70} \left(\frac{\lambda[X]}{d_{mx}[X]} \right) \right)^2$ using equation (4)

18: Compute $G_{e(dB)}[X] = G_{mx(dB)}[X] - L_\theta[X]$ using equation (5)

19: Compute $d_e[X] = \lambda[X] \sqrt{\frac{10 \left(\frac{G_{e(dB)}[X]}{10} \right)}{(\pi^2) \eta[X]}}$ using equation (7)

20: Compute $\Delta d[X] = d_{mx}[X] - d_e[X]$ using equation (8)

21: Output $f[X]$, $\lambda[n_x]$, $\eta[X]$, $d_{mx}[X]$, $\theta[X]$, $G_{mx(dB)}[n_x]$, $\theta_{3dB}[n_x]$, $L_\theta[X]$, $G_{e(dB)}[n_x]$, $d_e[n_x]$, $\Delta d[n_x]$

22: Next For Loop

23: End For Loop

24: End Module 2 ParAnal_using_f()

Algorithm 4 The procedure for the Module 3 ParAnal_using_η ()

Algorithm 4 The procedure for the Module 3 ParAnal_using_η ()

Module 3 ParAnal_using_η ()

- 1: Input η_L // lower value of efficiency
- 2: Input η_H // higher value of efficiency
- 3: Input n_x // number of steps to cover the range η_L to η_H
- 4: Declare the following arrays $\lambda[n_x], G_{mx(dB)}[n_x], \theta_{3dB}[n_x], L_\theta[X], G_{e(dB)}[n_x], d_e[n_x], \Delta d[n_x]$
- 5: Input λ
- 6: Input d_{mx}
- 7: Input θ
- 8: Declare the following arrays $\eta[n_x], d_{mx}[n_x], \theta[n_x]$
- 9: For $X = 1, X (n_x, X = X + 1$
 - 10: $\eta[X] = X \left(\frac{\eta_H - \eta_L}{n_x} \right)$
 - 11: $\lambda[X] = \lambda$
 - 12: $d_{mx}[X] = d_{mx}$
 - 13: $\theta[X] = \theta$
 - 14: Compute $G_{mx(dB)}[X] = 10 \text{Log} \left((\pi^2) \eta[X] \left(\frac{d_{mx}[X]}{\lambda} \right)^2 \right)$ using equation (1)
 - 15: Compute $\theta_{3dB}[X] = 70 \left(\frac{\lambda[X]}{d_{mx}[X]} \right)$ using equation (3)
 - 16: Compute $L_\theta[X] = 12 \left(\frac{\theta[X]}{70} \left(\frac{\lambda[X]}{d_{mx}[X]} \right) \right)^2$ using equation (4)
 - 17: Compute $G_{e(dB)}[X] = G_{mx(dB)}[X] - L_\theta[X]$ using equation (5)
 - 18: Compute $d_e[X] = \lambda[X] \sqrt{\frac{10 \left(\frac{G_{e(dB)}[X]}{10} \right)}{(\pi^2) \eta[X]}}$ using equation (7)
 - 19: Compute $\Delta d[X] = d_{mx}[X] - d_e[X]$ using equation (8)
 - 20: Output $\lambda[n_x], \eta[X], d_{mx}[X], \theta[X], G_{mx(dB)}[n_x], \theta_{3dB}[n_x], L_\theta[X], G_{e(dB)}[n_x], d_e[n_x], \Delta d[n_x]$
 - 21: Next For Loop
 - 22: End For Loop
 - 23: End Module 3 ParAnal_using_η ()

Algorithm 5 The procedure for the Module 4 ParAnal_using_ d_{mx} ()

Algorithm 5 The procedure for the Module 4 ParAnal_using_ d_{mx} ()

Module 4 ParAnal_using_ d_{mx} ()

- 1: Input d_{mxL} // lower value of antenna aperture diameter
- 2: Input d_{mxH} // higher value of antenna aperture diameter
- 3: Input n_x // number of steps to cover the range d_{mxL} to d_{mxH}
- 4: Declare the following arrays $\lambda[n_x], G_{mx(dB)}[n_x], \theta_{3dB}[n_x], L_\theta[X], G_{e(dB)}[n_x], d_e[n_x], \Delta d[n_x]$
- 5: Input λ
- 6: Input η
- 7: Input θ
- 8: Declare the following arrays $\eta[n_x], d_{mx}[n_x], \theta[n_x]$
- 9: For $X = 1, X (n_x, X = X + 1$

10: $d_{mx}[X] = X \left(\frac{d_{mxH} - d_{mxL}}{n_x} \right)$
 11: $\lambda[X] = \lambda$
 12: $\eta[X] = \eta$
 13: $\theta[X] = \theta$
 14: Compute $G_{mx(dB)}[X] = 10 \text{Log} \left((\pi^2) \eta[X] \left(\frac{d_{mx}[X]}{\lambda} \right)^2 \right)$ using equation (1)
 15: Compute $\theta_{3dB}[X] = 70 \left(\frac{\lambda[X]}{d_{mx}[X]} \right)$ using equation (3)
 16: Compute $L_\theta [X] = 12 \left(\frac{\theta[X]}{70} \left(\frac{\lambda[X]}{d_{mx}[X]} \right) \right)^2$ using equation (4)
 17: Compute $G_{e(dB)}[X] = G_{mx(dB)}[X] - L_\theta [X]$ using equation (5)
 18: Compute $d_e[X] = \lambda[X] \sqrt{\frac{10 \left(\frac{G_{e(dB)}[X]}{10} \right)}{(\pi^2) \eta[X]}}$ using equation (7)
 19: Compute $\Delta d[X] = d_{mx}[X] - d_e[X]$ using equation (8)
 20: Output $\lambda[n_x], \eta[X], d_{mx}[X], \theta[X], G_{mx(dB)}[n_x], \theta_{3dB}[n_x], L_\theta [X], G_{e(dB)}[n_x], d_e[n_x], \Delta d[n_x]$
 21: Next For Loop
 22: End For Loop
 23: End Module 4 ParAnal_using_ d_{mx} ()

Algorithm 6 The procedure for the Module 5 ParAnal_using_ θ ()

Algorithm 6 The procedure for the Module 5 ParAnal_using_ θ ()

Module 5 ParAnal_using_ θ ()
 1: Input θ_L // lower value of antenna pointing error
 2: Input θ_H // higher value of antenna pointing error
 3: Input n_x // number of steps to cover the range θ_L to θ_H
 4: Declare the following arrays $\lambda[n_x], G_{mx(dB)}[n_x], \theta_{3dB}[n_x], L_\theta [X], G_{e(dB)}[n_x], d_e[n_x], \Delta d[n_x]$
 5: Input λ
 6: Input η
 7: Input d_{mx}
 8: Declare the following arrays $\eta[n_x], d_{mx}[n_x], \theta[n_x]$
 9: For $X = 1, X (n_x, X = X + 1$
 10: $\theta[X] = X \left(\frac{\theta_H - \theta_L}{n_x} \right)$
 11: $\lambda[X] = \lambda$
 12: $\eta[X] = \eta$
 13: $d_{mx}[X] = d_{mx}$
 14: Compute $G_{mx(dB)}[X] = 10 \text{Log} \left((\pi^2) \eta[X] \left(\frac{d_{mx}[X]}{\lambda} \right)^2 \right)$ using equation (1)
 15: Compute $\theta_{3dB}[X] = 70 \left(\frac{\lambda[X]}{d_{mx}[X]} \right)$ using equation (3)
 16: Compute $L_\theta [X] = 12 \left(\frac{\theta[X]}{70} \left(\frac{\lambda[X]}{d_{mx}[X]} \right) \right)^2$ using equation (4)
 17: Compute $G_{e(dB)}[X] = G_{mx(dB)}[X] - L_\theta [X]$ using equation (5)
 18: Compute $d_e[X] = \lambda[X] \sqrt{\frac{10 \left(\frac{G_{e(dB)}[X]}{10} \right)}{(\pi^2) \eta[X]}}$ using equation (7)
 19: Compute $\Delta d[X] = d_{mx}[X] - d_e[X]$ using equation (8)
 20: Output $\lambda[n_x], \eta[X], d_{mx}[X], \theta[X], G_{mx(dB)}[n_x], \theta_{3dB}[n_x], L_\theta [X], G_{e(dB)}[n_x], d_e[n_x], \Delta d[n_x]$
 21: Next For Loop

22: End For Loop
 23: End Module 5 ParAnal_using_θ()

3. Results and discussion

The program was developed using Visual Basic for Applications that runs under Microsoft Excel platform. The program was used to conduct analysis on the impact of variations in one parameter of the antenna on another. The results of the parametric analysis of the antenna loss versus antenna pointing error are shown in Table 1 and Figure 1, where $\eta = 0.65$, $d_{mx} = 1.2$ m and θ (deg) = 0.5° . The results show that the antenna loss increases with frequency and the increase is quadratic in respect of the frequency in MHz, as shown in Figure 1.

The results of the parametric analysis of the antenna gain with and without pointing loss versus frequency are shown in Table 2 and Figure 2, where $\eta = 0.65$, $d_{mx} = 1.2$ m and θ (deg) = 0.5° . The results show that the antenna gain without pointing loss, G_{mx} (dB) and effective antenna gain with pointing loss, G_e (dB) increase logarithmically with frequency but the increase in G_{mx} (dB) is higher than

that of G_e (dB). As such, as frequency increases, the difference between G_{mx} (dB) and G_e (dB) increases.

The results of the parametric analysis of the effective antenna diameter with pointing loss, d_e (m) versus frequency are shown in Table 3 and Figure 3, where $\eta = 0.65$, $d_{mx} = 1.2$ m and θ (deg) = 0.5° . The results show that d_e (m) decreases quadratically with frequency. As such, as frequency increases, the difference between d_{mx} (m) and d_e (dB) increases.

The results of the parametric analysis of the antenna beam width, θ_{3dB} versus frequency are shown in Table 4 and Figure 4, where $\eta = 0.65$, $d_{mx} = 1.2$ m and θ (deg) = 0.5° . The results show that θ_{3dB} decreases inversely with frequency. As such, as frequency increases, the θ_{3dB} increases

Table 1 The results of parametric analysis of the antenna loss versus antenna pointing error, where $\eta = 0.65$, $d_{mx} = 1.2$ m and θ (deg) = 0.5°

S/N	f (MHz)	Antenna Pointing Loss, L_θ (dB)	S/N	f (MHz)	Antenna Pointing Loss, L_θ (dB)
1	1000	0.010	11	11000	1.185
2	2000	0.039	12	12000	1.411
3	3000	0.088	13	13000	1.656
4	4000	0.157	14	14000	1.920
5	5000	0.245	15	15000	2.204
6	6000	0.353	16	16000	2.508
7	7000	0.480	17	17000	2.831
8	8000	0.627	18	18000	3.174
9	9000	0.793	19	19000	3.536
10	10000	0.980	20	20000	3.918
11	11000	1.185	21	21000	4.320

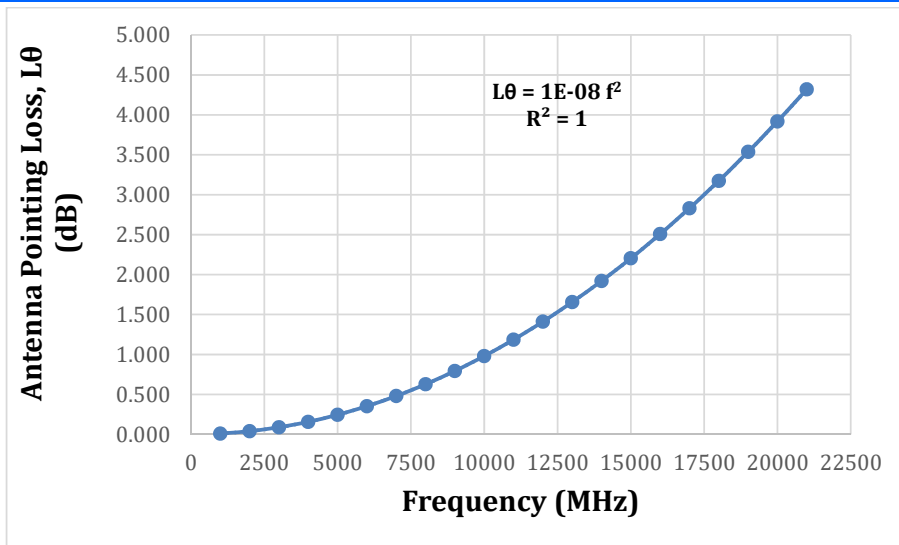


Figure 1 The graph of parametric analysis of the antenna loss versus antenna pointing error, where $\eta = 0.65$, $d_{mx} = 1.2$ m and θ (deg) = 0.5°

Table 2 The results of parametric analysis of the antenna gain with and without pointing loss versus frequency, where $\eta = 0.65$, $d_{mx} = 1.2$ m and θ (deg) = 0.5°

S/N	f (MHz)	Maximum Antenna Gain Without Pointing Loss, G _{mx} (dB)	Effective antenna gain with pointing loss, G _e (dB)	S/N	f (MHz)	Maximum Antenna Gain Without Pointing Loss, G _{mx} (dB)	Effective antenna gain with pointing loss, G _e (dB)
1	1000	20.113	20.104	11	11000	40.941	39.756
2	2000	26.134	26.095	12	12000	41.697	40.286
3	3000	29.656	29.568	13	13000	42.392	40.737
4	4000	32.155	31.998	14	14000	43.036	41.116
5	5000	34.093	33.848	15	15000	43.635	41.431
6	6000	35.676	35.324	16	16000	44.196	41.688
7	7000	37.015	36.535	17	17000	44.722	41.891
8	8000	38.175	37.548	18	18000	45.219	42.045
9	9000	39.198	38.405	19	19000	45.688	42.152
10	10000	40.113	39.134	20	20000	46.134	42.216
11	11000	40.941	39.756	21	21000	46.558	42.238

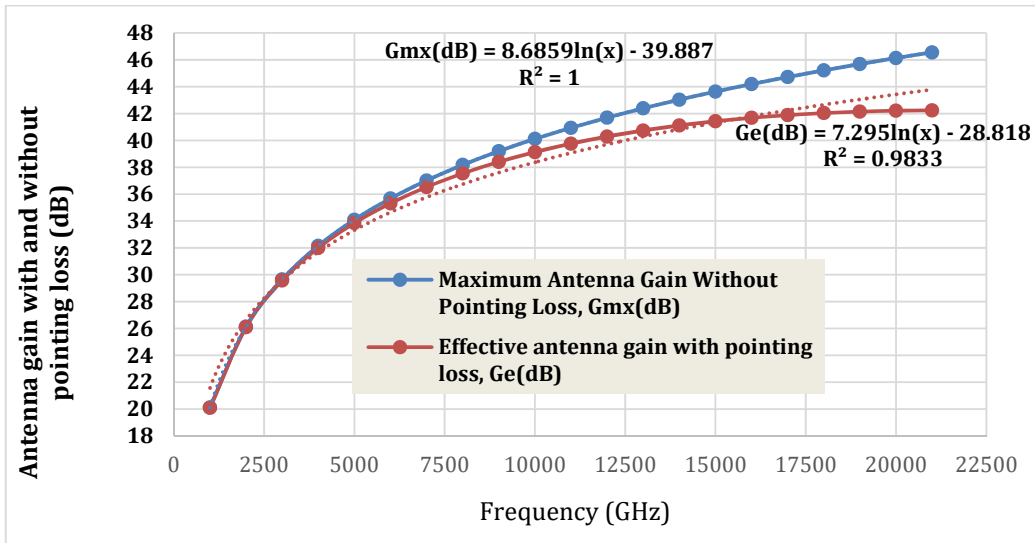


Figure 2 The graph of parametric analysis of the antenna gain with and without pointing loss versus frequency, where $\eta = 0.65$, $d_{mx} = 1.2$ m and θ (deg) = 0.5°

Table 3 The results of parametric analysis of the effective antenna diameter with pointing loss versus frequency, where $\eta = 0.65$, $d_{mx} = 1.2$ m and θ (deg) = 0.5°

S/N	f (MHz)	Actual antenna diameter without pointing loss, d_{mx} (m)	Effective antenna diameter with pointing loss, d_e (m)	S/N	f (MHz)	Actual antenna diameter without pointing loss, d_{mx} (m)	Effective antenna diameter with pointing loss, d_e (m)
1	1000	1.2	1.199	11	11000	1.2	1.047
2	2000	1.2	1.195	12	12000	1.2	1.02
3	3000	1.2	1.188	13	13000	1.2	0.992
4	4000	1.2	1.179	14	14000	1.2	0.962
5	5000	1.2	1.167	15	15000	1.2	0.931
6	6000	1.2	1.152	16	16000	1.2	0.899
7	7000	1.2	1.135	17	17000	1.2	0.866
8	8000	1.2	1.116	18	18000	1.2	0.833
9	9000	1.2	1.095	19	19000	1.2	0.799
10	10000	1.2	1.072	20	20000	1.2	0.764
11	11000	1.2	1.047	21	21000	1.2	0.73

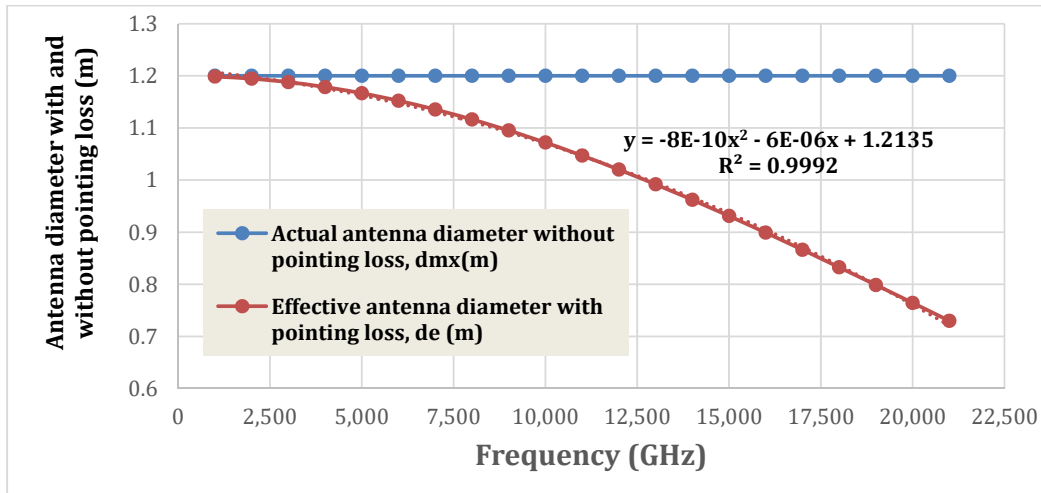


Figure 3 The graph of parametric analysis of the effective antenna diameter with pointing loss versus frequency, where $\eta = 0.65$

Table 4 The results of parametric analysis of the antenna beam width, θ_{3dB} versus frequency, where $\eta = 0.65$, $d_{mx} = 1.2$ m and θ (deg) = 0.5°

S/N	f (MHz)	Antenna beam width, θ_{3dB}	S/N	f (MHz)	Antenna beam width, θ_{3dB}
1	1000	17.500	11	11000	1.591
2	2000	8.750	12	12000	1.458
3	3000	5.833	13	13000	1.346
4	4000	4.375	14	14000	1.250
5	5000	3.500	15	15000	1.167
6	6000	2.917	16	16000	1.094
7	7000	2.500	17	17000	1.029
8	8000	2.188	18	18000	0.972
9	9000	1.944	19	19000	0.921
10	10000	1.750	20	20000	0.875
11	11000	1.591	21	21000	0.833

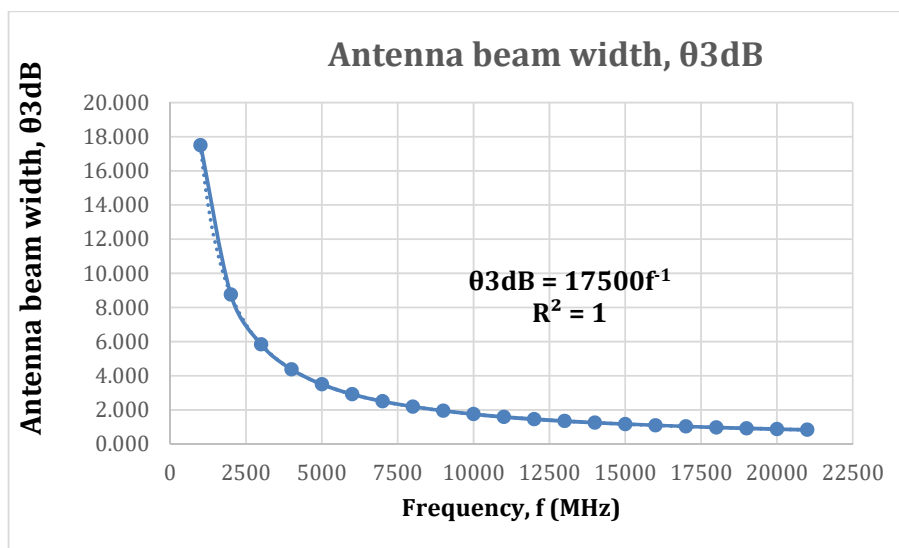


Figure 4 The graph of parametric analysis of the the antenna beam width, θ_{3dB} versus frequency, where $\eta = 0.65$, $d_{mx} = 1.2$ m and θ (deg) = 0.5°

The results of the parametric analysis of the antenna pointing loss, $L\theta$ versus antenna pointing error shown in Table 5 and Figure 5, where $\eta = 0.65$, $d_{mx} = 1.2$ m and $f = 2000$ MHz. The results show that the antenna pointing loss, $L\theta$ increases quadratically with antenna pointing error, θ . As such, as θ increases, the antenna pointing loss, $L\theta$ increases quadratically with θ in degrees.

The results of the parametric analysis of the effective antenna gain, G_e (dB) versus antenna pointing error are shown in Table 6 and Figure 6, where $\eta = 0.65$, $d_{mx} = 1.2$ m and $f = 2000$ MHz. The results show that G_e (dB) decreases quadratically with frequency. As such, as θ

increases, the effective antenna gain, G_e (dB) decreases quadratically with θ in degrees.

The results of the parametric analysis of the effective antenna diameter, d_e (m) versus antenna pointing error shown in Table 7 and Figure 7, where $\eta = 0.65$, $d_{mx} = 1.2$ m and $f = 2000$ MHz. The results show that the effective antenna diameter, d_e (m) decreases with third power of antenna pointing error, θ . As such, as θ increases, the effective antenna diameter, d_e (m) decreases with θ raised to power of 3.

Table 5 The parametric analysis of the antenna pointing loss versus antenna pointing error where $\eta = 0.65$, $d_{mx} = 1.2$ m and $f = 2000$ MHz

S/N	Antenna pointing error, θ (deg)	Antenna loss, $L\theta$ (dB)	S/N	Antenna pointing error, θ (deg)	Antenna loss, $L\theta$ (dB)
1	0	0	11	0.5	3.918
2	0.05	0.039	12	0.55	4.741
3	0.1	0.157	13	0.6	5.642
4	0.15	0.353	14	0.65	6.622
5	0.2	0.627	15	0.7	7.68
6	0.25	0.98	16	0.75	8.816
7	0.3	1.411	17	0.8	10.031
8	0.35	1.92	18	0.85	11.324
9	0.4	2.508	19	0.9	12.696
10	0.45	3.174	20	0.95	14.145
11	0.5	3.918	21	1	15.673

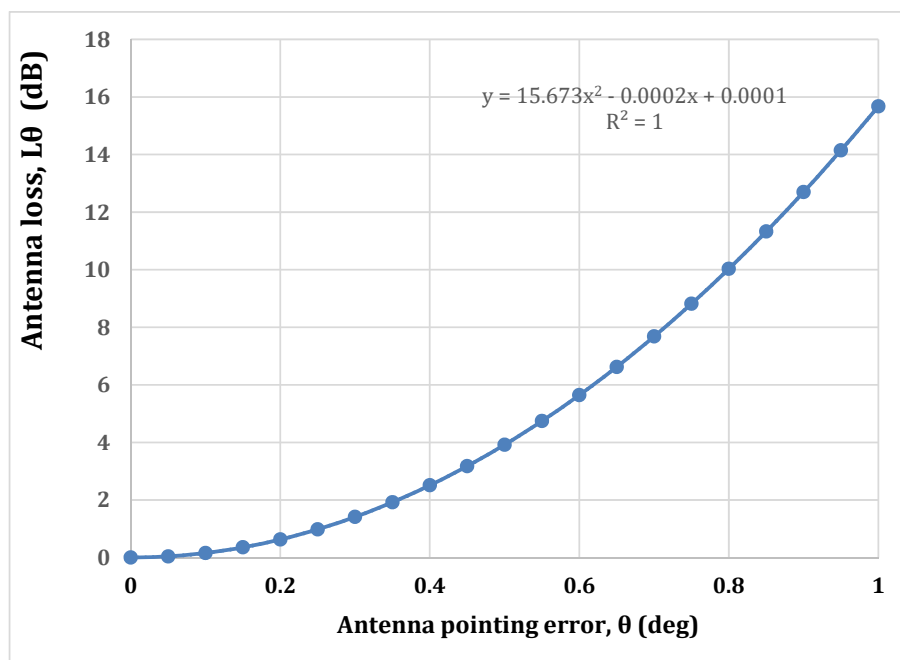


Figure 5 The graph of parametric analysis of the antenna pointing loss versus antenna pointing error where $\eta = 0.65$, $d_{mx} = 1.2$ m and $f = 2000$ MHz

Table 6 The parametric analysis of the effective antenna gain, Ge(dB) versus antenna pointing error where $\eta = 0.65$, $d_{mx} = 1.2$ m and $f = 2000$ MHz

S/N	Antenna pointing error, θ (deg)	Effective antenna gain with pointing loss, Ge(dB)	S/N	Antenna pointing error, θ (deg)	Effective antenna gain with pointing loss, Ge(dB)
1	0	46.134	11	0.5	42.216
2	0.05	46.095	12	0.55	41.393
3	0.1	45.977	13	0.6	40.491
4	0.15	45.781	14	0.65	39.512
5	0.2	45.507	15	0.7	38.454
6	0.25	45.154	16	0.75	37.318
7	0.3	44.723	17	0.8	36.103
8	0.35	44.214	18	0.85	34.81
9	0.4	43.626	19	0.9	33.438
10	0.45	42.96	20	0.95	31.989
11	0.5	42.216	21	1	30.46

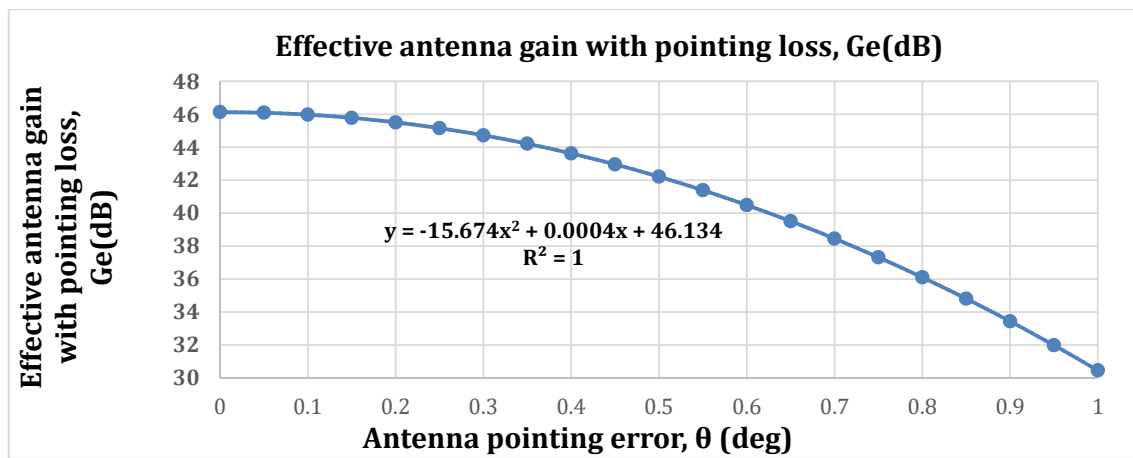


Figure 6 The graph of parametric analysis of the effective antenna gain, Ge(dB) versus antenna pointing error where $\eta = 0.65$, $d_{mx} = 1.2$ m and $f = 2000$ MHz

Table 7 The parametric analysis of the effective antenna diameter, de(m) versus antenna pointing error where $\eta = 0.65$, $d_{mx} = 1.2$ m and $f = 2000$ MHz

S/N	Antenna pointing error, θ (deg)	Effective antenna diameter with pointing loss, de (m)	S/N	Antenna pointing error, θ (deg)	Effective antenna diameter with pointing loss, de (m)
1	0	1.2	11	0.5	1.2
2	0.05	1.195	12	0.55	1.195
3	0.1	1.179	13	0.6	1.179
4	0.15	1.152	14	0.65	1.152
5	0.2	1.116	15	0.7	1.116
6	0.25	1.072	16	0.75	1.072
7	0.3	1.02	17	0.8	1.02
8	0.35	0.962	18	0.85	0.962
9	0.4	0.899	19	0.9	0.899
10	0.45	0.833	20	0.95	0.833
11	0.5	0.764	21	1	0.764

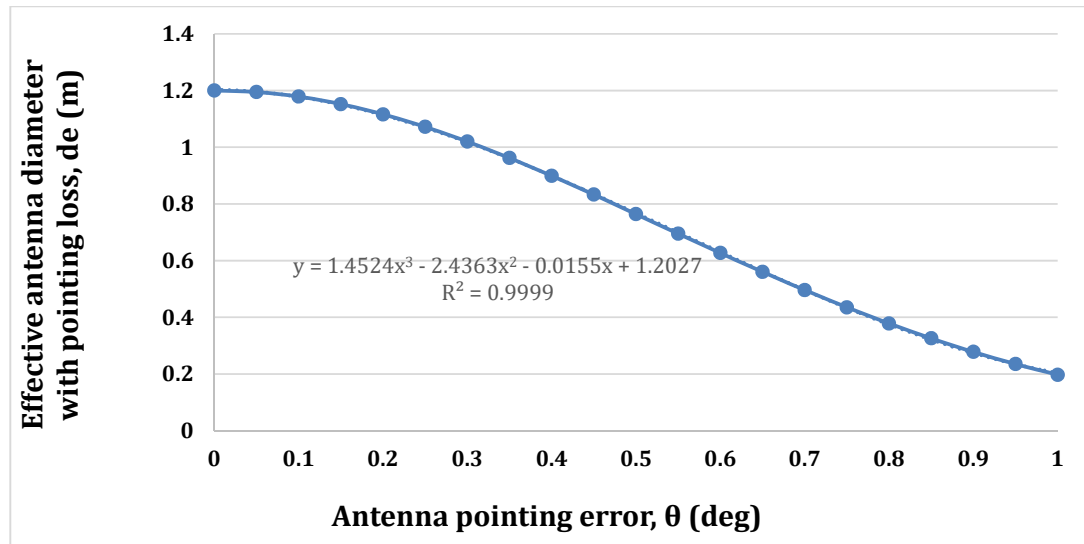


Figure 7 The graph of parametric analysis of the effective antenna diameter, d_e (m) versus antenna pointing error θ where $\theta = 0.65$, $d_{mx} = 1.2$ m and $f = 2000$ MHz

4. Conclusion

The program for parametric analysis of antenna pointing error impact on parabolic antenna gain, effective antenna diameter, and antenna pointing and beam width is presented. The program also captures the effect of frequency for a given antenna pointing error of the listed parameters. The program was implemented in Visual Basic for Application. Sample numerical examples were used to demonstrate the applicability of the program in conducting such parametric analysis.

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